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Application Number:	61990354
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Confirmation Number:	7828
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First Named Inventor/Applicant Name:	Erik Norvell
Customer Number:	27045
Filer:	Steven Ware Smith/Kara Coffman
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Harmonic-Noise like Audio Signal Classifier (Discriminator)

1 BACKGROUND

1.1 Technical background / Existing technology

Modern audio codecs consists of multiple compression schemes optimized for signals with different properties. With practically no exception, speech-like signals are processed with time-domain codec, while music signals are processed with transform-domain codec. Coding schemes that are supposed to handle both speech and music signals require a mechanism to recognize input signal (speech-music classifier) and switch between the appropriate codec modes. An overview illustration of a multimode audio codec using mode decision logic based on the input signal is shown in Figure 1.

In a similar manner among the class of music signals one can discriminate more noise like music signals and harmonic music signals, and build classifier and optimal coding scheme for each of these groups. This abstraction of creating a classifier to determine the class of a signal which then controls the mode decision is illustrated in Figure 2. These implementations deal with finding a better classifier for discrimination of harmonic and noise like music signals.

1.2 Problems with existing solutions

Variety of speech-music classifiers are used in the field of audio coding. However, these classifiers cannot discriminate between different classes in the space of music signals. Many classifiers do not provide enough resolution to discriminate between classes which are needed in a complex multimode codec.

2 BRIEF SUMMARY OF THE PROPOSED SOLUTION / ABSTRACT

The problem of harmonic and noise-like music segments discrimination is solved by a novel metric, calculated directly on the frequency-domain coefficients. The metric is based on the distribution of pre-selected spectral peaks candidates and the average peak-to-noise floor ratio.

3 ADVANTAGES OF THE PROPOSED SOLUTION

The proposed solution allows harmonic and noise-like music segments to be identified, which in turn allows for optimal coding of these signal types as illustrated in figure 3. This coding concept provides a superior quality over the conventional coding schemes.

4 DETAILED DESCRIPTION OF THE PROPOSED SOLUTION AND FIGURES

The exemplary audio codec, and therefore the presented audio discriminator/classifier, operate on short blocks (e.g. 20ms) of the input waveform. The presented embodiment show exemplary numerical values which are preferred for the embodiment at hand. It should be understood that these numerical values are given only as examples and may be adapted to the audio codec at hand.

Before performing spectral analysis, the input waveform may optionally be pre-emphasized with a first-order high-pass filter $H(z) = 1 - 0.68z^{-1}$

The discrete Fourier transform (DFT) is used to convert the signal into the frequency domain. The spectral analysis is performed once per frame using 256-point fast Fourier transform (FFT).

An FFT is performed on the windowed input signal to obtain one set of spectral parameters:

$$X(k) = \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}$$

Where $k = 0, \dots, 255$, is an index of frequency coefficients, and n is an index of waveform samples. It should be noted that any length N of the transform may be used.

The discriminator requires knowledge of spectral peaks location. Spectral peaks are defined as coefficients with absolute value above an adaptive threshold, which is based on the ratio of peak and noise-floor envelopes.

The noise-floor estimation algorithm operates on the absolute values of transform coefficients $|X(k)|$. Instantaneous noise-floor energies $E_{nf}(k)$ are estimated according to the recursion:

$$E_{nf}(k) = \alpha E_{nf}(k-1) + (1 - \alpha) |X(k)|^2$$

$$\alpha = \begin{cases} 0.9978 & \text{if } |X(k)|^2 \geq E_{nf}(k-1) \\ 0.6472 & \text{if } |X(k)|^2 \leq E_{nf}(k-1) \end{cases}$$

The particular form of the weighting factor α minimizes the effect of high-energy transform coefficients and emphasizes the contribution of low-energy coefficients. Finally the noise-floor level E_{nf} is estimated by simply averaging the instantaneous energies $E_{nf}(k)$.

$$E_{nf} = \sum_{k=0}^{255} E_{nf}(k)$$

The peak-picking algorithm requires knowledge of noise-floor level and average level of spectral peaks. The peak energy estimation algorithm is similar to the noise-floor estimation algorithm, but instead of low-energy, it tracks high-spectral energies.

$$E_p(n) = \beta E_p(n-1) + (1 - \beta) |X(n)|^2$$

$$\beta = \begin{cases} 0.4223 & \text{if } |X(n)|^2 > E_p(n-1) \\ 0.8929 & \text{if } |X(n)|^2 \leq E_p(n-1) \end{cases}$$

In this case the weighting factor β minimizes the effect of low-energy transform coefficients and emphasizes the contribution of high-energy coefficients. The overall peak energy \bar{E}_p is estimated by simply averaging the instantaneous energies.

$$\bar{E}_p = \frac{1}{N} \sum_{n=0}^{N-1} E_p(n)$$

When the peak and noise-floor levels are calculated, a threshold level τ is formed,

$$\tau = \left(\frac{\bar{E}_p}{\bar{E}_n} \right)^\gamma \bar{E}_n$$

with $\gamma = 0.8837$. Transform coefficients are compared to the threshold, and the ones with amplitude above it, form a vector of peak candidates.

An alternative threshold value, which may require less computational complexity, is to use the instantaneous peak envelope level with a fixed scaling factor

$$\theta(n) = E_p(n) \cdot 0.64$$

The peak candidates will then be all the coefficients with the squared amplitude above the instantaneous threshold level

$$\begin{cases} |X(n)|^2 > \theta(n), n \in \mathcal{P} \\ |X(n)|^2 \leq \theta(n), n \notin \mathcal{P} \end{cases}$$

where \mathcal{P} denotes the frequency ordered set of positions of index candidates. Considering the FFT spectrum, some peaks will be broad and consist of several transform coefficients, while others are narrow and are represented by a single coefficient. To get a representation of individual coefficients, consecutive peak candidates are assumed to be part of a broader peak. By finding the maximum squared amplitude $|X(n)|^2$ of the transform coefficients in a range of consecutive peak positions $n-k-1 \leq n \leq k$, a refined set \mathcal{P} is created where the broad peaks are represented by the maximum position in each range. Figure 4 illustrates the derivation of the peak envelope and noise floor, and the peak selection algorithm.

The above calculations serve to generate features that are used for forming the classifier decision: an estimate of the peak scarcity \mathcal{S} and a peak-to-noise floor ratio $\mathcal{P/N}$. The peak scarcity may be represented using the average distance \bar{d}_i between peaks

$$S = \left\{ \frac{P_{\text{ref}}}{P_{\text{ref}} + P_{\text{ref}}} \cdot \frac{P_{\text{ref}}}{P_{\text{ref}} + P_{\text{ref}}} \cdot \frac{P_{\text{ref}}}{P_{\text{ref}} + P_{\text{ref}}} \right\}$$

where N_{ref} is the number of refined peaks in the set P . The P_{ref} may be calculated as

$$P_{\text{ref}} = \frac{P}{N_{\text{ref}}}$$

The classifier decision may be formed using these features in combination with a decision threshold. We can name these decisions

$$issparse = S > S_{\text{thr}}$$

$$isclean = PNR > PNR_{\text{thr}}$$

The outcome of these decisions may be used to form different classes of signals. An illustration of these classes is shown in Figure 6. Since the classification is based on two binary decisions, the total number of classes may be at most 4. As a next step, the codec decision can be formed using the class information, which is illustrated in Table 1.

Table 1: Possible classes formed using two feature decisions.

	<i>isclean</i>	<i>Issparse</i>
Class A	false	False
Class B	true	False
Class C	true	True
Class D	false	true

In the following step in the audio codec, the decision is to be made which processing steps to apply to which class. This mapping will depend on the characteristics and capabilities of the different coding modes or processing steps available. As an example, perhaps Codec mode 1 would handle Class A and Class C, while Codec mode 2 would handle Class B and Class D. The coding mode decision can be the final output of the classifier to guide the encoding process. The coding mode decision would typically be transferred in the bitstream together with the codec parameters from the chosen coding mode.

It should be understood that the above classes may be further combined with other classifier decisions. The combination may result in a larger number of classes, or they may be combined using a priority order such that the presented classifier may be overruled by another classifier, or vice versa that the presented classifier may overrule another classifier.

The described implementation is a high-resolution music type discriminator with application audio coding. The decision logic of the discriminator is based on statistics of positional distribution of frequency coefficients with prominent energy.

5 GENERAL

An overview of the discriminator can be seen in Figures 3 and 7.

Thus, Figure 3 is a schematic block diagram of a discriminator according to embodiments. The discriminator comprises:

- an input unit (not shown) configured to receive an input signal representing an audio signal to be handled,
- a Framing unit for producing frames of the input signal,
- an optional Pre-emphasis unit for filtering a frame of the input signal with a high-pass filter,
- a Frequency transforming unit for performing a Fourier transform on the output of the Pre-emphasis unit (or the output of the Framing unit if the Pre-emphasis unit is omitted) to obtain frequency coefficients,
- a Peak/Noise envelope analysis unit for performing an envelope analysis on the frequency coefficients to calculate a threshold value,
- a Peak candidate selection unit for selecting peak candidates from among the frequency coefficients based on the threshold value,
- a Peak candidate refinement unit for refining the selection of peak candidates to account for broad peaks in the Fourier transform spectrum to produce a set of selected peaks,
- a Feature calculation unit for generating features (e.g., peak scarcity and peak-to-noise floor ratio PNR) based on the set of selected peaks,
- a Class decision unit for making a classification decision based on the features,
- a Coding mode decision unit for determining a coding mode based on the classification decision,
- a Multi-mode encoder unit for encoding frames of the audio signal based on the determined coding mode,
- a Bit-streaming/Storage unit for streaming and/or storing the encoded frames, and
- an output unit (not shown) for outputting the encoded frames of the audio signal.

All these units could be implemented in hardware. There are numerous variants of circuitry elements that can be used and combined to achieve the functions of the units of the encoder. Such variants are encompassed by the embodiments. Particular examples of hardware implementation of the discriminator are implementation in digital signal processor (DSP) hardware and integrated circuit technology, including both general-purpose electronic circuitry and application-specific circuitry.

The discriminator described herein could alternatively be implemented e.g. by one or more of a processor and adequate software with suitable storage or memory therefore, in order to perform the discriminatory action on an input signal, according to the embodiments described herein, see Fig. 7. The incoming signal is received by an input (IN), to which the processor and the memory are connected, and the discriminatory representation of an audio signal (parameters) obtained from the software is outputted from the output (OUT).

The technology described above may be used e.g. in a sender, which can be used in a mobile device (e.g. mobile phone, laptop) or a stationary device, such as a personal computer.

It is to be understood that the choice of interacting units or modules, as well as the naming of the units are only for exemplary purpose, and may be configured in a plurality of alternative ways in order to be able to execute the disclosed process actions.

It should also be noted that the units or modules described in this disclosure are to be regarded as logical entities and not with necessity as separate physical entities. It will be appreciated that the scope of the technology disclosed herein fully encompasses other embodiments which may become obvious to those skilled in the art, and that the scope of this disclosure is accordingly not to be limited.

Reference to an element in the singular is not intended to mean "one and only one" unless explicitly so stated, but rather "one or more." All structural and functional equivalents to the elements of the above-described embodiments that are known to those of ordinary skill in the art are expressly incorporated herein by reference and are intended to be encompassed hereby. Moreover, it is not necessary for a device or method to address each and every problem sought to be solved by the technology disclosed herein, for it to be encompassed hereby.

In the preceding description, for purposes of explanation and not limitation, specific details are set forth such as particular architectures, interfaces, techniques, etc. in order to provide a thorough understanding of the disclosed technology. However, it will be apparent to those skilled in the art that the disclosed technology may be practiced in other embodiments and/or combinations of embodiments that depart from these specific details. That is, those skilled in the art will be able to devise various arrangements which, although not explicitly described or shown herein, embody the principles of the disclosed technology. In some instances, detailed descriptions of well-known devices, circuits, and methods are omitted so as not to obscure the description of the disclosed technology with unnecessary detail. All statements herein reciting principles, aspects, and embodiments of the disclosed technology, as well as specific examples thereof, are intended to encompass both structural and functional equivalents thereof. Additionally, it is intended that such equivalents include both currently known equivalents as well as equivalents developed in the future, e.g. any elements developed that perform the same function, regardless of structure.

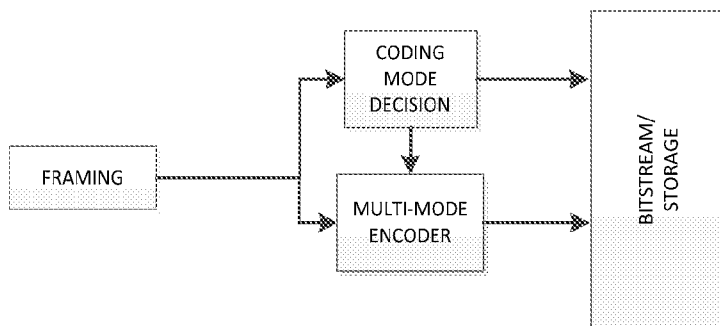
Thus, for example, it will be appreciated by those skilled in the art that the figures herein can represent conceptual views of illustrative circuitry or other functional units embodying the principles of the technology, and/or various processes which may be substantially represented in computer readable medium and executed by a computer or processor, even though such computer or processor may not be explicitly shown in the figures.

The functions of the various elements including functional blocks may be provided through the use of hardware such as circuit hardware and/or hardware capable of executing software in the form of coded instructions stored on computer readable medium. Thus, such functions and illustrated functional blocks are to be understood as being either hardware-implemented and/or computer-implemented, and thus machine-implemented.

The embodiments described above are to be understood as a few illustrative examples of the present invention. It will be understood by those skilled in the art that various modifications, combinations and changes may be made to the embodiments without departing from the scope of the present invention. In particular, different part solutions in the different embodiments can be combined in other configurations, where technically possible.

FIGURE 1

Signal discriminator in an audio codec. Usage of the discriminators output to switch between different coding modes.



PROVISIONAL APPLICATION FOR PATENT COVER SHEET – Page 1 of 2

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Given Name (first and middle [if any])	Family Name or Surname	Residence (City and either State or Foreign Country)
Erik	Norvell	Stockholm, Sweden
Volodya	Grancharov	Solna, Sweden
Additional inventors are being named on the _____ separately numbered sheets attached hereto		
TITLE OF THE INVENTION (500 characters max):		
Harmonic-Noise like Audio Signal Classifier (Discriminator)		
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<input checked="" type="checkbox"/> The address corresponding to Customer Number:	27045	
OR		
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ENCLOSED APPLICATION PARTS (check all that apply)		
<input type="checkbox"/> Application Data Sheet. See 37 CFR 1.76	<input type="checkbox"/> CD(s), Number of CDs _____	
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<input checked="" type="checkbox"/> Specification (e.g. description of the invention) <i>Number of Pages</i> <u>11</u>		
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SIGNATURE /Ronald J. Ward, Reg No 54,870/

Date May 8, 2014

TYPED or PRINTED NAME Ronald J. Ward

REGISTRATION NO. 54,870
(if appropriate)

TELEPHONE 972-583-8656

Docket Number: **P43568-US1**

6 ABBREVIATIONS

<u>Abbreviation</u>	<u>Explanation</u>
DFT	Discrete Fourier Transform
FFT	Fast Fourier Transform
MDCT	Modified Discrete Cosine Transform
ACELP	Code-Excited Linear Prediction
VQ	Vector Quantization
PNR	Peak to Noise floor ratio

FIGURE 2

Signal discriminator in an audio codec, on the path higher level discriminator (speech/music discriminator)

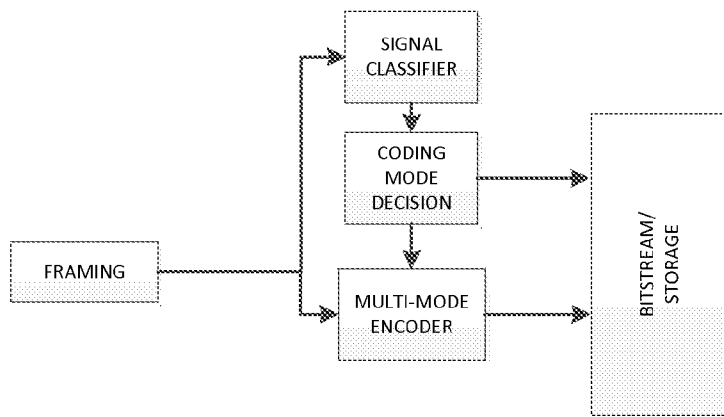


FIGURE 3

Block diagram of the proposed signal discriminator, including the following modules: framing, pre-emphasis, frequency transform, calculating threshold for spectral peaks, calculating metric on the positional distribution of spectral peaks, decision module

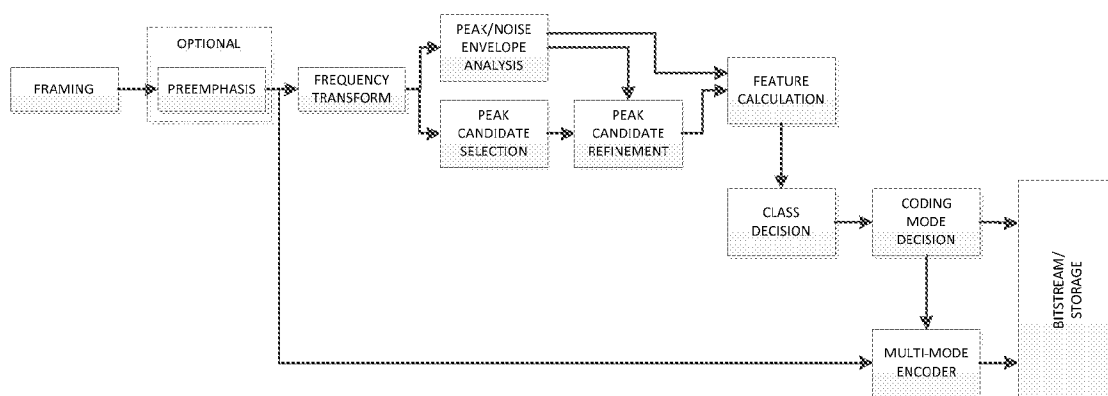


FIGURE 4

Illustration of peak selection algorithm and instantaneous

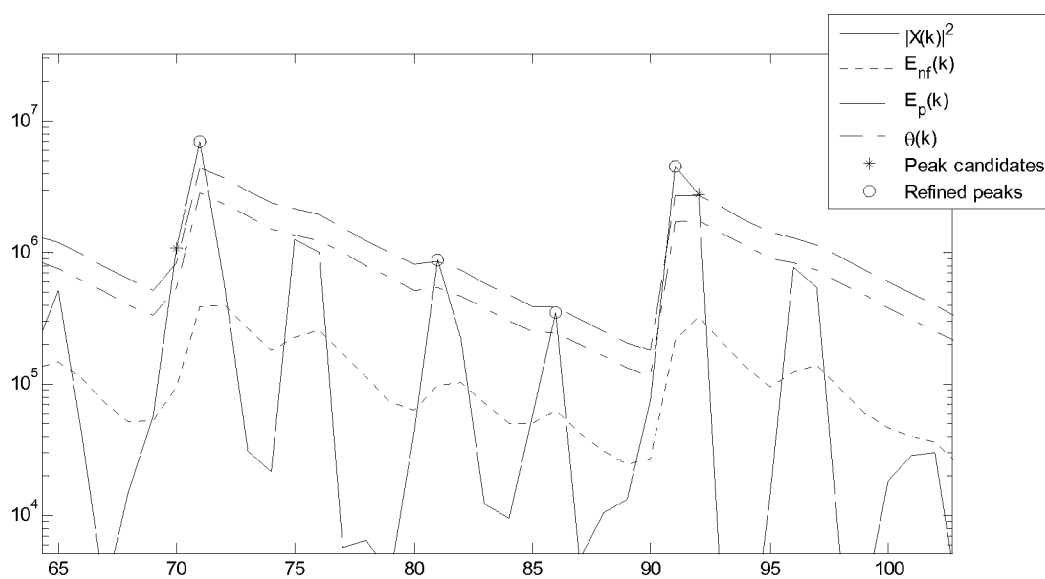


FIGURE 5

Illustration of peak distances d_i

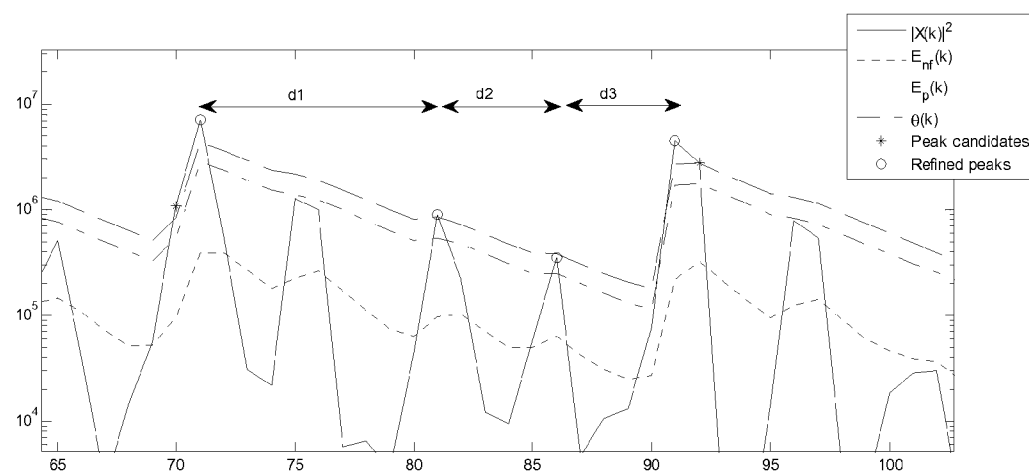
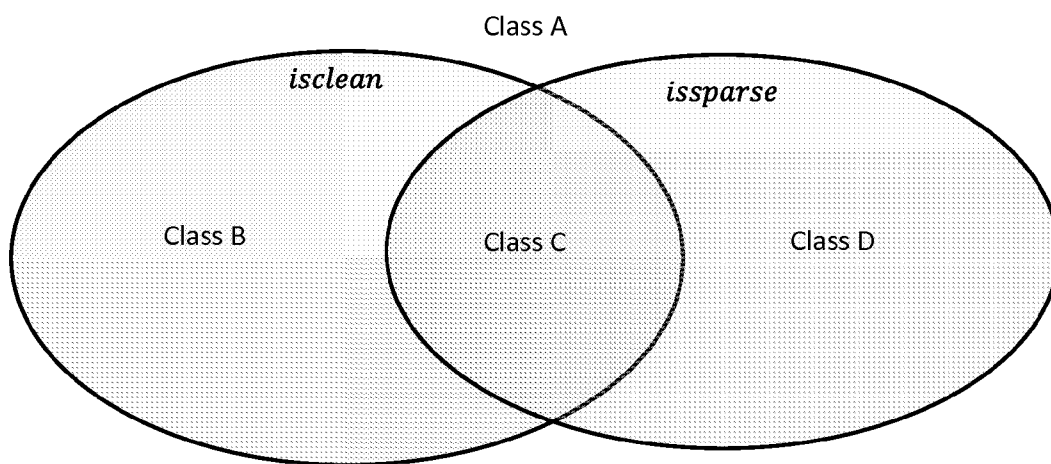


FIGURE 6
Example of class decisions as a Venn diagram

$$\begin{aligned} \text{issparse} &= S > S_{THR} \\ \text{isclean} &= PNR > PNR_{THR} \end{aligned}$$



Discriminator

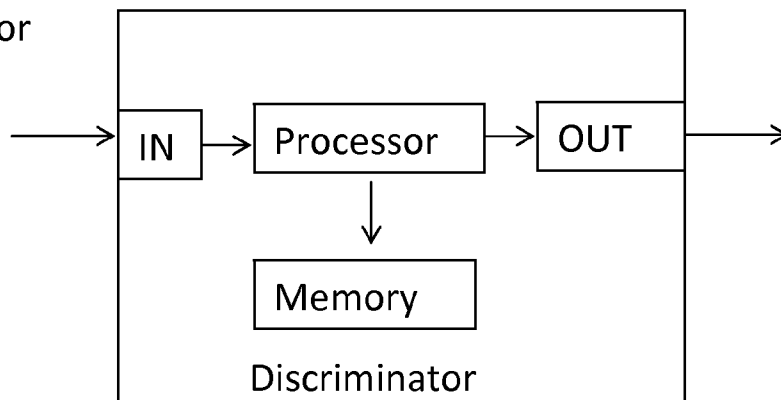


Figure. 7